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A TIME DOMAIN CHARACTERIZATION OF POSITIVE-REAL MATRICES

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# A TIME DOMAIN CHARACTERIZATION OF POSITIVE-REAL MATRICES

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### ABSTRACT

The following time-domain characterization of any positive-real matrix is established: Let  $\underline{w}(t)$  be an n n matrix distribution and let  $\underline{W}(s)$  be its Laplace transform. Necessary and sufficient conditions for  $\underline{W}(s)$  to be a positive-real matrix are the following.

1. 
$$\underline{w}(t) = \underline{A} \mathcal{E}^{(i)}(t) + \underline{w}_{o}(t)$$

Here, A is a real symmetric nonnegative-definite constant matrix and  $w_0(t)$  is a real matrix distribution of zero order, whose support is contained in 0.44  $\infty$ .

2. Let

$$w_h(t) = \frac{1}{2} [w(t) + w^T(-t)].$$

For every  $n \times 1$  constant vector y the quantity  $y * y_h(t) y$  is a non-negative-definite distribution.

1. We continue in this report the discussion of time-domain characterization of positive-real matrices  $\mathbb{W}(s)$  that was presented in a previous report [1]. The present development is considerably briefer and yet more general since the assumption of rationality for  $\mathbb{W}(s)$  is no longer imposed. Indeed, a complete time-domain characterization of any arbitrary positive-real matrix, whether rational or not, is obtained and the conclusions of our previous report [1] are included as a special case.

Our development is based upon a time-domain representation for the unit impulse response matrix for a single-valued linear time-invariant continuous, causal, and passive n-port, which was presented in a previous paper [2]. An analysis for one-ports that overlaps the present work (i.e., it achieves the necessity of condition 1 of the forthcoming theorem for the case where n=1) has been given by Konig and Meixner [3,4].

We shall continue to use the notations that were employed in [1]. Moreover, we especially call the reader's attention to Sec. II of [1] wherein a summary is given of certain fundamental concepts and theorems that will be used in the subsequent discussion. This summary will not be repeated here. No other material in [1] other than Sec. II is needed for an understanding of this report.

Another concept that we shall make use of here is that of the order of a distribution. The order of a matrix distribution is that least nonnegative integer r for which the (r + 2) th - order primitives of the elements in the matrix are all continuous functions.

 $Q_{\delta}(t)$  is of zero order whereas  $A_{\delta}(t)$  (t) is of first order.

The third term on the right-hand side of (3) is a symbolic expression for that matrix distribution which assigns to each function  $\phi$  in D the number

$$= u(t) \int_{-\infty}^{\infty} dK(\eta) \int_{0}^{\infty} \phi^{(2)}(t) \frac{1-\cos \eta t}{\eta^{2}} (1+\eta^{2}) dt$$

$$= u(t) \int_{-\infty}^{\infty} \cos \eta t dK(\eta) \cdot \phi(t) + u(t) \int_{-\infty}^{\infty} (1-\cos \eta t) dK(\eta) \cdot \phi^{(2)}(t)$$

$$(4)$$

(See  $\begin{bmatrix} 2 \end{bmatrix}$ ; theorem  $4 \end{bmatrix}$ ). Since the elements of

$$u(t) \int_{-\infty}^{\infty} \cos \eta t \, d \, K(\eta)$$
 (5)

are all locally integrable functions of t, (5) is a zero-order matrix distribution. Also, the elements of

$$u(t)$$
  $(1-\cos\eta t) dK(\eta)$ 

are continuous functions everywhere and the second term in (4) defines the second distributional derivative of (6). Therefore, (6) is also a zero-order matrix distribution.

The fourth term on the right-hand side of (3) is a symbolic expression that assigns to each function  $\phi$  in D the number,

$$\underbrace{\mathcal{L}}_{\infty}^{\infty} d\underline{L}(\eta) \underbrace{\int_{0}^{\infty} \phi^{(1)}(t) \frac{1-\cos\eta t}{\eta}}_{\eta} dt + \underbrace{\int_{0}^{\infty} \phi^{(2)}(t) \sin\eta t}_{\eta} dt \Big)$$

$$= -u(t) \underbrace{\int_{0}^{\infty} \sin\eta t}_{\eta} d\underline{L}(\eta) \cdot \phi(t) + u(t) \underbrace{\int_{0}^{\infty} \sin\eta t}_{\eta} d\underline{L}(\eta) \cdot \phi^{(2)}(t).$$

$$(7)$$

(Again, see [2; theorem 4]). Now, the elements of

$$u(t) \stackrel{\sim}{\underset{\sim}{\sum}} \sin \eta t \, dL(\eta) \tag{8}$$

are all continuous functions for all t and it follows that both terms on the right-hand side of (7) define zero-order matrix distributions.

This establishes all of condition 1.

3.

Constructing  $w_h(t)$  according to (2) and using the properties of the matrices Q, A,  $K(\eta)$ , and  $L(\eta)$ , we see that

$$\mathbf{W}_{h}(t) = \frac{1}{2} \mathbf{J}^{\circ \bullet} e^{j\eta t} (1 + \eta^{2}) d[\mathbf{K}(\eta) + j\mathbf{L}(\eta)].$$

Therefore,

$$y^* w_h(t) y = \frac{1}{2} \int_{-\infty}^{\infty} e^{j\eta t} (1 + \eta^2) d \left\{ y^* \left[ \underline{K}(\eta) + j\underline{L}(\eta) \underline{y} \right] \right\}.$$

Since  $y * [K(\eta) - jL(\eta)] y$  is a real nondecreasing bounded function of  $\eta$  for all choices of y,  $y * [K(\eta) + jL(\eta)] y$  must have the same properties. Condition 2 now follows from the Bochner-Schwartz theorem. (See Sec. II of [1] for a statement of this theorem.)

Sufficiency: Assume that w(t) satisfies conditions 1 and 2. Since  $y_{w_h}(t)y$  is nonnegative-definite, it is also of slow growth [5: Vol. II, p. 132]. Moreover, in forming  $w_h(t)$  from w(t), only those terms in the elements of w(t) that are concentrated on the origin can cancel out. These terms are known to be only the delta function and a finite number of derivatives, which are also distributions of slow growth. It follows that w(t) is a matrix distribution of slow growth. Therefore, the Laplace transform w(t) of w(t) exists and is analytic for Re s > 0. In addition, since w(t) is real, w(t) is real for real positive s.

The proof will be completed when we show that

$$y * W_h(s) y \ge 0 \tag{9}$$

for Re s > 0 and for every choice of y, where

$$W_{h}(s) = \frac{1}{2} \left[ W(s) + W^{*}(s) \right]. \tag{10}$$

According to the distributional inverse Laplace transformation, we may relate w(t) to w(c+jw) for c>0 through the symbolic expression,

$$e^{-ct}w(t) = \frac{1}{2\pi} \underbrace{\mathcal{L}^{\infty}_{\infty} \mathcal{W}(c + j\omega)}_{e^{j\omega t}} d\omega$$
 (11)

Since W(s) is real for real positive s, the reflection principle shows that  $W(c-j\omega) = \overline{W(c+j\omega)}$ . Using this fact, we may write

$$e^{-ct} w(t) + e^{ct} w^{T}(-t) = \frac{1}{\pi} v^{\infty} w_{h} (c+j\omega) e^{j\omega t} d\omega$$

and

$$y^* \left[ e^{-ct} \underline{w}(t) + e^{ct} \underline{w}^T(-t) \right] \underline{y} = \frac{1}{\pi} \sum_{\alpha} y^* \underline{w}_{\alpha} (c + j\omega) y e^{j\omega t} d\omega \qquad (c)$$

If we can show that the left-hand side of (12) is a nonnegative-definite distribution, then (9) will follow from the Bochner-Schwartz theorem and the fact that  $\mathbb{V}(s)$  is analytic for Re s > 0.

Now 9

$$e^{-ct}AS^{(1)}+e^{ct}A^{T}S^{(1)}(-t) = 2Ac\delta(t)$$

and, consequently, we obtain from the decomposition (1) that

$$e^{-ct}w(t) + e^{ct}w^{T}(-t) = 2Acc(t) + e^{-ct}w_{O}(t) + e^{ct}w_{O}^{T}(-t)$$

Moreover, (1) also shows that

$$2w_h(t) = w_0(t) + w_0^T(-t).$$
 (13)

Since  $w_0(t)$  is of zero order,  $w_0(t)$ ,  $\phi(t)$  depends only on the values of  $\phi(t)$  over  $0 \le t < \infty$  and not on the derivatives of  $\phi(t)$  [5: Vol. I, P. 93]. It follows that

$$e^{-ct}w_0(t) = e^{-c|t|}w_0(t)$$
.

Therefore,

$$y^* \left[ e^{-ct} w(t) + e^{ct} w^T(-t) \right] y = 2y^* Ayc c(t) + e^{-c} \left[ t \right] y^* \left[ w_0(t) + w_0^T(-t) \right] y^{(14)}$$

Now,  $\chi^*A\chi \geq 0$  and g(t) is a nonnegative-definite distribution. Therefore, the first term on the right-hand side of (14) is nonnegative-definite for every y and for c>0.

Also, by condition 2 and (13),  $y*[w](t)+w^T(-t)]y$  is nonnegative-definite for every y. It is a fact that  $e^{-c|t|}(c>0)$  is also nonnegative-definite. We now invoke a theorem of Schwartz [5: Vol. II, p. 134, theorem XIX] to conclude that the second term on the right-hand side of (14) is nonnegative-definite for every y and for c>0. Q.E.D.

The principal conclusion of our previous report [1], which characterizes the unit impulse response matrix of any lumped linear fixed finite and passive n-port, now appears as a special case of the above theorem.

Corollary: Let w(t) be an  $n \times n$  matrix distribution and let w(s) be its Laplace transform. Necessary and sufficient conditions for w(s) to be positive-real and rational are the following.

1. 
$$w(t) = A s^{(1)}(t) + B s(t) + w_1(t)$$
.

Here, A is a real symmetric nonnegative-definite matrix, B is a real constant matrix, and w1(t) is a real matrix distribution whose elements consist of finite linear combinations of terms of the form,

$$u(t)t^{\nu}e^{-\nu t}, \qquad (15)$$

where  $\nu$  is a nonnegative integer and  $\gamma$  is a complex constant.

2. For every  $n \times 1$  constant vector y and for  $w_h(t)$  given by (2) the quantity y \* w (t) y is a nonnegative-definite distribution.

When dealing with one-ports our conclusions simplify into the following.

Corollary: Let w(t) be a distribution and let the function W(s) be its Laplace transform. W(s) is positive-real if and only if

1. 
$$w(t) = A S^{(i)}(t) + w_0(t)$$
,

where A is a nonnegative constant and  $w_0(t)$  is a real distribution of zero order with support in  $0 \le t \le \infty$ , and

2. the even part,

$$w_{b}(t) = \frac{1}{2} [w(t) + w(-t)],$$
 (16)

of w(t) is a nonnegative-definite distribution.

Corollary: Let w(t) be a distribution and let the function W(s) be its Laplace transform. W(s) is positive-real and rational if and only if

1. 
$$w(t) = A s^{(1)}(t) + B s^{(t)} + w_{1}(t)$$
,

where A and B are real constants with A ≥ 0 and w,(t) is a real distribution that consists of a finite linear combination of terms of the form (15), and

2. the even part (16) of w(t) is a nonnegative-definite distribution.